

# JEPPIAAR INSTITUTE OF TECHNOLOGY

"Self-Belief | Self Discipline | Self Respect"



DEPARTMENT

OF

# **ELECTRONICS & COMMUNICATION ENGINEERING**

# LECTURE NOTES EC8553/ DISCRETE-TIME SIGNAL PROCESSING (Regulation 2017)

Year/Semester: III/V ECE 2021 – 2022

Prepared by Mrs.S.Mary Cynthia Assistant Professor/ECE

# **OBJECTIVES:**

EC8553

- To learn discrete fourier transform, properties of DFT and its application to linear filtering
- To understand the characteristics of digital filters, design digital IIR and FIR filters and apply these filters to filter undesirable signals in various frequency bands

DISCRETE-TIME SIGNAL PROCESSING

- To understand the effects of finite precision representation on digital filters
- To understand the fundamental concepts of multi rate signal processing and its applications
- To introduce the concepts of adaptive filters and its application to communication engineering. •

## **UNIT I - DISCRETE FOURIER TRANSFORM**

Review of signals and systems, concept of frequency in discrete-time signals, summary of analysis & synthesis equations for FT & DTFT, frequency domain sampling, and Discrete Fourier transform (DFT) - deriving DFT from DTFT, properties of DFT - periodicity, symmetry, circular convolution. Linear filtering using DFT. Filtering long data sequences - overlap save and overlap add method. Fast computation of DFT - Radix-2 Decimation-in-time (DIT) Fast Fourier transform (FFT), Decimation-in-frequency (DIF) Fast Fourier transform (FFT). Linear filtering using FFT.

# **UNIT II - INFINITE IMPULSE RESPONSE FILTERS**

Characteristics of practical frequency selective filters. Characteristics of commonly used analog filters -Butterworth filters, Chebyshev filters. Design of IIR filters from analog filters (LPF, HPF, BPF, BRF) -Approximation of derivatives, Impulse invariance method, Bilinear transformation. Frequency transformation in the analog domain. Structure of IIR filter - direct form I, direct form II, Cascade, parallel realizations.

# **UNIT III - FINITE IMPULSE RESPONSE FILTERS**

Design of FIR filters - symmetric and Anti-symmetric FIR filters - design of linear phase FIR filters using Fourier series method - FIR filter design using windows (Rectangular, Hamming and Hanning window), Frequency sampling method. FIR filter structures - linear phase structure, direct form realizations.

### UNIT IV FINITE WORD LENGTH EFFECTS

Fixed point and floating point number representation - ADC - quantization - truncation and rounding quantization noise - input / output quantization - coefficient quantization error - product quantization error overflow error - limit cycle oscillations due to product quantization and summation - scaling to prevent overflow.

# UNIT V INTRODUCTION TO DIGITAL SIGNAL PROCESSORS

DSP functionalities - circular buffering - DSP architecture - Fixed and Floating point architecture principles -Programming – Application examples.

## **OUTCOMES:**

After studying this course, the student should be able to:

- Apply DFT for the analysis of digital signals and systems
- Design IIR and FIR filters
- Characterize the effects of finite precision representation on digital filters •
- Design multirate filters •
- Apply adaptive filters appropriately in communication systems.

### **TEXT BOOKS:**

1. John G. Proakis & Dimitris G.Manolakis, -Digital Signal Processing - Principles, Algorithms & Applications, Fourth Edition, Pearson Education / Prentice Hall, 2007. (UNIT I – V)

# REFERENCES

1 Emmanuel C. Ifeachor & Barrie. W. Jervis, -Digital Signal Processingl, Second Edition, Pearson Education / Prentice Hall, 2002.

2. A. V. Oppenheim, R.W. Schafer and J.R. Buck, -Discrete-Time Signal Processingl, 8th Indian Reprint, Pearson, 2004.

3. Sanjit K. Mitra, —Digital Signal Processing – A Computer Based Approach<sup>||</sup>, Tata Mc Graw Hill, 2007.

4. Andreas Antoniou, -Digital Signal Processing, Tata Mc Graw Hill, 2006.

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**TOTAL: 60 PERIODS** 

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EC 8553 Discrete - Time Signal Processing. Onit - TDiscrete Fourier Transform Syllabus:-Review of signals and systems, Concept of frequency in discreto-time signals, Summary of analysis & synthesis equations for FT & DTFT, Frequency domain Sempling, Discrete - Fourier-Transform (DFT) - deriving DFT from DTFT, Properties of DFT - periodicity, Symmetry, Circular Convolution, Linear filtering using DFT-Filtoring long data sequences - overlap save and overlap add method, Fast Computation of DFT, Rodix - 2 Decimation - is - time (DIT), Fast Fourier transform (IFT), Decimation - in - fraquency (DDF) Linear filtering wing FFT. Review of Signals and Systems:-Signal - A signal is defined as the physical quantity that varies with time, space or

any other independent variables.

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Continuous time Signals (CT):

The signals that are defined for every instant of time are known as continuous-time Signals. They are denoted by xcl). Example:x(t) Discrete-time signal (DT):-The signals that are defined at discrete-instant of time are known as discretetime signals They are continuous is amplitude and discrete-in-time. They are denoted by x(n) System:-It is defined as a physical device that generates a response er output Signal, for a given input signal.  $\mathcal{D}\mathcal{T}$ x(n) DT Syste x(t) System ) y(t) y(n) = T 「たいの〕 Y(t) = T[x(t)] yet) is the transformed form of r(t). yen) is the transformed form of x(n).

Discrete Fourier) Transform: It is defined as.  $\chi(k) = \sum_{n=0}^{N-1} \chi(n) e^{-j2 \prod nk} N$ where N-> lengte of sequence n= 0, 1, .... N-1 < = 0, 1, ... N-1 EX. 1. 1. Compute DFT 2(1) = { 1, 0, 1, 0} x(n)={1,0,1,03  $\chi(\kappa) = \leq \chi(n)e^{-j2\pi nk}$  $n = 0, 1, 2, \dots N-1 \implies n = 0, 1, 2, 3.$ N = 4,K=0,1,2,...N-1=) K=0,1,2,3  $\mathcal{K}(k) = \sum_{n=0}^{4-1} \chi(n) e^{-j\frac{\pi n}{4}} \frac{\pi}{4}$  $\chi(\kappa) = \sum_{n=0}^{3} \chi(n) e^{-j \frac{\prod n K}{2}} \qquad K = 0, 1, \dots, N-1$   $(\kappa) = \chi(0) e^{-j \frac{\prod \chi 0 \chi k}{2}} + \chi(0) e^{-j \frac{\prod K}{2}} + \chi(2) e^{-j \frac{1}{2}}$ x(3)e-j311k Y(K)= 1×1+0+1.0+1.0+0 Civon,  $\gamma(0) = 1$  $\chi(z) = 1$   $\chi(1x) = 1 + e^{-j\pi x}$  $\chi(z) = 0$ ,  $\sqrt{1}$  $\chi(0) = 1 + e^{-j\pi 0} = 1 + 1 = \pi(0) = 2$  $\chi(1) = 1 + e^{-j \Pi x} = \chi(1) = 1 - 1 = \chi(1) = 0$ k=0, K=1, = 1+ (0517- SinTi)

$$\begin{aligned} & k = 2, \\ & \chi_{(2)} = 1 + e^{-j2\pi} \\ & \eta_{(3)} = 1 + (\cos \sqrt{n}) - j\sin \sqrt{n} + 1 - 0 \\ & k = 3, \\ & \chi_{(3)} = 1 + e^{-j3\pi} \\ & = 1 + (\cos \sqrt{n}) - j\sin \sqrt{n} \\ & = 1 + (-1 - 0) \\ & \chi_{(3)} = 0 \\ & \chi_{(3)} = \frac{1}{2} \\ & \chi_{($$

2 npm

$$\begin{aligned} k=1, & \chi(t):=e^{-jT/2} + 2e^{-jT(t)} + 3e^{-j2T/2} \\ &= -j + 2(\cos T t - j\sin T t) + 3(\cos 3T/2 - j\sin 3T/2) \\ &= -j + 2(\cos T t - j\sin T t) + 3(\cos 3T/2 - j\sin 3T/2) \\ &= -j + 2(-1) + 3j \\ &= -j - 2 + 3j \\ \hline \chi(t) = -2 + 2j \\ \chi(t) = e^{-jT} + 2e^{-j\pi T} + 3e^{-j\pi T} \\ &= -1 + 2(-1) + 3(-1) \\ &= -1 + 2 - 3 \\ \hline \chi(t) = -2 \\ k=3, & \chi(t) = e^{-jST/2} + 2e^{-jT^{-3}} + 3e^{-jT^{-3}} \\ &= -j + 2(-1) + 3(-j) \\ \hline \chi(t) = -2 \\ \chi(t) =$$

Step 1: (yro) >



Outer circle -> Anti clochwise Inner circle-s clochwise.

changing is anticlack -wise



Step 3 :- y(2)



Y(&) = & - &+3 y(&)= 3 y(1) = 0 + 0 - 3 + 8 - 1 $y(1)^{2} - 2$ 

step 4: 4(3)



step 5: 4(4)





K=0, X(0)=1+1+2+) X107=5  $\chi(1) = 1 + e^{-j\pi/2} + e^{-j\pi} + e^{-j\pi/2}$ kal, = 1 + (-j) + a(-i) + (-j)= | - [ - 2 + ]  $\chi(1) = -1$  $\chi(a) = 1 + e^{-j\pi} + ae^{-j\pi/2} + e^{-j\pi/2}$ k=a,  $= 1 + (-1) + 2(1) + (05 \pi (6/a) - jsin \pi (6/a)$ = | - | + 2 - ] k=3,  $\chi(3)=1+e^{-\int \pi^3/a}+ae^{-j3\pi}-j\pi^{9/2}$ = 1 - j + 2(-1) + (-j)= (-) - 2+)  $\left[\chi(3)=-1\right]$ x(K)= \$5,-1,1,-1}  $h(n) = \{1, 0, 3, 4\}$  $H(k) = \sum_{n=0}^{N-1} h(n) e^{-j \vartheta T T n k / N}$  $h^{=0} = \frac{4-1}{h(R)} = \frac{1}{2} = \frac{1}{h(R)} = \frac{1}{2} = \frac{1}{2}$ 

H(k): 
$$\int_{h=0}^{3} h(h)e^{-j\pi hk} z$$
  
H(k):  $h(0)x_{1} + h(1)e^{-j\pi k} = \frac{1}{4} + \frac{$ 

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$$\begin{aligned} \chi(k) &= \{5, -1, 1, -3\} \\ h(k) &= \{10, -20 + 2j, -20, -20 - 2ij\} \\ Y(k) &= \chi(k) + h(k) \\ Y(k) &= \{5, -1, 1, -1\} + \{10, -20 + 2j, -2, -20 - 2ij\} \\ DFT \quad \text{Sequence is,} \\ Y(k) &= \{50, 2 - 2i, -20, 20 + 2ij\} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{Nm}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} = \frac{2}{N} \quad Y(k) e^{j\frac{N}{N}} \\ y(n) &= \frac{1}{N} \quad Y($$

$$= \frac{1}{4} \left\{ \frac{1}{80} + \frac{1}{8} - \frac{1}{2} \right\}^{3} - \frac{1}{8} (-1) + (\frac{1}{8} + \frac{1}{2})(-j) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} \right\}$$

$$\frac{1}{9(1) = 14}$$

$$h=0, \quad y(a) = \frac{1}{4} \left\{ \frac{1}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} e^{-j\pi a} + (\frac{1}{8} + \frac{1}{8}) e^{-j\pi a} \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} (-1) + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} e^{-j\pi a} + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} e^{-j\pi a} + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} e^{-j\pi a} + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} e^{-j\pi a} + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} \right\}^{-1} + (\frac{1}{8} + \frac{1}{8})(-1) \right\}$$

$$= \frac{1}{4} \left\{ \frac{5}{80} + (\frac{1}{8} - \frac{1}{8}) \right\}^{-1} - \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} \right\}^{-1} + \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} + \frac{1}{8} \right\}^{-1} + \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} + \frac{1}{8} \right\}^{-1} + \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} + \frac{1}{8} + \frac{1}{8} \right\}^{-1} + \frac{1}{8} \left\{ \frac{1}{8} + \frac{1}{8} + \frac{1}{8} + \frac{1}{8} \right\}^{-1} + \frac{1}{8} \left\{ \frac{1}{8} + \frac{1$$

Convolution: Formula: y(n) = x(n) \* h(n) where y(n) - Output, x(n) - Isput, h(n) - Impulse Response In Frequency domain, Y(K) = X(K) H(K)

Linear Convolution: EX.1.5. Determine Output of the system  $\mathcal{H}(n) = \{1, -1, 2, -2\}$ and  $\mathcal{H}(n) = \{1, 3, 2, 1\}$  using linear/ tabular Convolution.

yin = xin \* hin

EX.1.6. Using linear convolution find  $\chi(n) = \chi_1(n) + \chi_2(n)$ and  $\chi_1(n) = \{1, 2, 3, 4\}$  as  $\chi_2(n) = \{5in -\frac{10}{4}\}$  if N = 4.

$$n = 0, 1, ... N-1$$

$$= 0, 1, 2, 3$$

$$\chi_{2}(0) = Sin0 = 0$$

$$\chi_{2}(1) = SinT_{4} = 0.707$$

$$\chi_{2}(3) = Sin2T_{4} = 1$$

$$\chi_{3}(3) = Sin3T_{4} = 0.707$$

$$\chi_{3}(4) = Sin3T_{4} = 0.707$$

$$\chi_{10}^{(1)} = \begin{pmatrix} 0 & 0.707 & 1 & 0.707 \\ 1 & 0 & 0.707 & 1 & 0.707 \\ 2 & 6 & 1.414 & 2 & 1.414 \\ 3 & 0 & 2.121 & 3 & 2.121 \\ 4 & 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 0 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 0 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ 0 & 2.828 & 4 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 4.828 & 7.042 & 6.121 & 8.828 \\ \chi_{10} = \begin{cases} 1 & 0.707 & 0.707 & 0.414 & 0.77 \\ 1 & 0.71 & 0 & -11 \\ 1 & 1 & 1 & 0 \\ 0 & -1 & 1 & 1 \\ 0 & -1 & 1 & 1 \\ 1 & 1 & 0 & -11 \\ 1 & 1 &$$

C.S.S.S. cross correlation:-It is denoted by  $\gamma_{\chi \ell} = \chi_1(l) + \chi_2(-l)$ EX.1.8. Find cross correlation of the given signals  $x_1(l) = \{1, 1, -1, 0\}$  and  $x_{\otimes}(l) = \{1, 3, 5, 7\}$ Criven:  $\chi_1(1) = \{1, 1, -1, 0\}$  $\chi_{g}(l) = \{1, 3, 5, 7\}$  $\gamma_{g}(-l) = \{7, 5, 3, 1\}$ YX= [ 7,12,1,-1,-2,-1,0] Linear Filtering (or) Filtering methods of DFT. 1. Overlap Add method -> using linear convolution 2. Overlap Sare metted > using matrix / Circular convolution. These methods used for low impulse signal to process high or long duration input sequence. EX. 1.9, perform linear convolution of finite duration Sequence h(n)={1,2}. I/P sequence x(n)={1,2,-1,2,3, -2,-3,-1,1,1,2,-13.  $h(n) = \{1, 2\}$  $\chi(n) = \{1, 2, -1, 2, 3, -2, -3, -1, 1, 1, 2, -1\}$ Cillon :

$$\chi_1(n) = \{1, 2\}; \quad \chi_0(n) = \{-1, 2\}; \quad \chi_3(n) = \{3, -2\};$$
  
 $\chi_4(n) = \{-3, -1\}; \quad \chi_5(n) = \{1, 1\}; \quad \chi_6(n) = \{2, -1\};$ 

step 1: 
$$h(n) = \{1, 2\}$$
  
 $Y_1(n) = \chi_1(n) * h(n)$   
 $Y_1(n) = \{1, a\} * \{1, a\}$ 

Wring linear Convolution.

Overlap Add metaod:

$$\begin{array}{c|c} h(n) & 1 & 2 \\ 1 & 1 & 2 \\ 2 & 2 & 4 \\ \end{array} \quad \begin{array}{c} y_1(n) = \begin{cases} 1, \ 4, \ 4 \\ \end{cases} \\ y_1(n) = \begin{cases} 1, \ 4, \ 4 \\ \end{cases}$$

step 2:-  

$$y_2(n) = \{-1, 2\} \times \{1, 2\}$$
  
 $y_2(n) = \{-1, 0, 4\}$   
 $y_2(n) = \{-1, 0, 4\}$   
 $2 = 2, 4, ...$ 

Step 3 :-  $y_3(n) = \{3, -23 + \{1, 2\}$   $x_3(n) = \{3, -23 + \{1, 2\}$   $x_3(n) = \{3, 4, -4\}$   $y_3(n) = \{3, 4, -4\}$   $y_4(n) = \{1, 2\} + \{-3, -1\}$  $y_4(n) = \{1, 2\} + \{-3, -1\}$ 

$$\chi_{4}(m) = \frac{1}{2}$$
  
-3 -3 -6  $\chi_{4}(m) = \frac{2}{3}$  -3 -7, -2  
-1 -1 -2

steps:-



Tabulation ?-



and services

Overlap\_Save Method :-

Wring Circular convolution, (NXM)-1 & "M is always 2". N- h(n) length.  $\chi(n) = \{1, 2, -1, Q, 3, -2, -3, -1, 1, 1, 2, -1\}$  $\chi_{4}(n) = \{1, 2, -1\}$   $\chi_{4}(n) = \{-3, -1, 1\}$  $X_{Q}(m) = \{-1, 2, 3\}$   $X_{5}(m) = \{1, 1, 2\}$  $\chi_3(n) = \{3, -2, -3\}$   $\chi_b(n) = \{2, -1, 0\}$ Step 1: y(n) = xr (n × h(n)  $y_{1}(n) = \begin{vmatrix} 1-2+0 \\ 2+2+0 \\ -1+4+0 \end{vmatrix} = \begin{vmatrix} 2-1 \\ 4 \\ 3 \end{vmatrix}$ y(n)={-1,4,3} step 2:- y(n) = x2(n) + h(n)  $Y_{a}(n) = \begin{bmatrix} -1 & 3 & 2 \\ 2 & -1 & 3 \\ 3 & 2 & -1 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 0 \end{bmatrix} = \begin{bmatrix} -1+6+0 \\ 2-2+0 \\ 3+4+0 \end{bmatrix} = \begin{bmatrix} 5 \\ 0 \\ 7 \end{bmatrix}$ Y2(n)=[5 0 7]  $step 3 = Y_3(n) = \chi_3(n) + h(n)$ 

$$y_3(n) = \left[ -3, 4, 7 \right]$$

Step 4:-

$$\begin{aligned} & \mathcal{Y}_{4}(\mu) = \mathcal{N}_{4}(n) \star h(n) \\ & \mathcal{Y}_{4}(n) = \begin{bmatrix} -3 & 1 & -1 \\ -1 & -3 & 1 \\ 1 & -1 & -3 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 0 \end{bmatrix} = \begin{bmatrix} -3 + 2 \\ -1 - 2 \end{bmatrix} = \begin{bmatrix} -1 \\ -7 \\ -1 \end{bmatrix} \\ & \mathcal{Y}_{4}(n) = \begin{bmatrix} -3 & 1 & -1 \\ 1 & -1 & -3 \end{bmatrix} \\ & \mathcal{Y}_{4}(n) = \begin{bmatrix} -1 & -7 & -1 \end{bmatrix} \\ & \mathcal{Y}_{5}(n) = \chi_{5}(n) \star h(n) \\ & \mathcal{Y}_{5}(n) = \begin{bmatrix} 1 & 2 & 1 \\ 1 & 1 & 2 \\ 2 & 1 & 1 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 0 \end{bmatrix} = \begin{bmatrix} 1 + 4 \\ 1 + 2 \\ 2 + 2 \end{bmatrix} = \begin{bmatrix} 5 \\ 3 \\ 4 \end{bmatrix} \\ & \mathcal{Y}_{5}(n) = \begin{bmatrix} 5 & 3 + 4 \end{bmatrix} \\ & \mathcal{Y}_{6}(n) = \begin{bmatrix} 5 & 3 + 4 \end{bmatrix} \\ & \mathcal{Y}_{6}(n) = \begin{bmatrix} 2 & 0 & -1 \\ -1 & 2 & 0 \\ 0 & -1 & 2 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 0 \end{bmatrix} = \begin{bmatrix} -1 + 4 \\ 0 - 2 \end{bmatrix} = \begin{bmatrix} 2 \\ 3 \\ -2 \end{bmatrix} \\ & \mathcal{Y}_{6}(n) = \begin{bmatrix} 2 & 0 & -1 \\ -1 + 4 \\ 0 - 2 \end{bmatrix} = \begin{bmatrix} 2 \\ 3 \\ -2 \end{bmatrix} \end{aligned}$$

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Proof 1- $X(K) = \bigotimes_{n=0}^{N-1} x(n) e^{j \operatorname{eTink}} N$   $X_{1}(K) = \bigotimes_{n=0}^{N-1} x_{1}(n) e^{-j \operatorname{eTink}} N$   $X_{0}(K) = \bigotimes_{n=0}^{N-1} \chi_{0}(n) e^{-j \operatorname{eTink}} N$   $X_{0}(K) = \bigotimes_{n=0}^{N-1} \chi_{0}(n) e^{-j \operatorname{eTink}} N$   $L.1+J \cdot \operatorname{operfa} x_{1}(n) + b \chi_{0}(n) = \bigotimes_{n=0}^{N-1} \chi(n) e^{-j \operatorname{eTink}} N$   $= \bigotimes_{n=0}^{N-1} (\alpha \pi_{1}(n) + b \chi_{0}(n)) e^{-j \operatorname{eTink}} N$   $= \bigotimes_{n=0}^{N-1} a \chi_{1}(n) e^{-j \operatorname{eTink}} N$   $= \sum_{n=0}^{N-1} a \chi_{1}(n) e^{-j \operatorname{eTink}} N$   $= \sum_{n=0}^{N-1} a \chi_{1}(n) e^{-j \operatorname{eTink}} N$   $= \sum_{n=0}^{N-1} a \chi_{1}(K) + b \chi_{0}(K)$ 

a) Time Shifting:  
If DFT [
$$\chi(n)$$
] =  $\chi(k)$  \_j  $\Im IIMk = N$   
Hen DFT [ $\chi(n-m)$ ] =  $\chi(k) e$   
 $\Im DFT [\chi(n+m)] = \chi(k) e$   
 $\Re DFT [\chi(n+m)] = \chi(k) e$   
 $\Re DFT [\chi(n+m)] = \chi(k) e^{-j \Im IInk/N}$   
 $\Im FT[\chi(n)] = \chi(k) = \sum_{n=0}^{N-1} \chi(n) e^{-j \Im IInk/N}$   
 $\Im FT[\chi(n-m)] = \sum_{n=0}^{N-1} \chi(n-m) e^{-j \Im IInk/N}$   
 $\bigwedge Rssume P = n-m \implies n = P+m$   
 $\bigwedge Rssume P = n-m \implies n = P+m$   
 $\lim_{n \to 0} P = 0 - DFT [\chi(n-m)] = \sum_{p=0}^{N-1} \chi(p) e^{-j \Im IInk/N}$   
 $\Im FT[\chi(n-m)] = \chi(k) e^{-j \Im IIPK/N} -j \Im IInk/N$   
 $\Im FT[\chi(n-m)] = \chi(k) e^{-j \Im IINK/N}$ 

3) Time Reversal:  
If DFT of 
$$X(n) = X(k)$$
  
 $+kan DFT of X(n-k) - X(N-k)$   
 $DFT of X(n) = X(k) = \sum_{n=0}^{N-1} X(n)e^{-jatink}$   
 $DFT of X(N-n) = \sum_{n=0}^{N-1} X(n-n)e^{-jatink}$   
 $m = N-n \implies n = N-m$   
 $DFT (X(N-n) = \sum_{m=0}^{N-1} X(m)e^{-jatink}$   
 $DFT (X(N-n)) = \sum_{m=0}^{N-1} X(m)e^{-jatink}$   
 $DFT [X(N-n)] = \sum_{m=0}^{N-1} X(m)e^{-jatink}$   
 $DFT [X(N-n)] = X(m)e^{-jatink}$   
 $DFT [X(N-n)] = X(N-K)$   
4) Parsevals Theorem:  
 $If DFT [Y(n) = X(k)$   
 $from \sum_{n=0}^{N-1} X(n)e^{-jx(k)} = X(k)$   
 $fhen \sum_{n=0}^{N-1} X(n)e^{-jx(k)}$   
 $fhen \sum_{n=0}^{N-1} X(n)e^{-jx(k)}$   
 $K(k) = \sum_{n=0}^{N-1} X(n)e^{-jx(k)}$   
 $K(k) = \sum_{n=0}^{N-1} X(n)e^{-jx(n)k}$   
 $E = \sum_{n=0}^{N-1} X(n)e^{-jx(n)k}$   
 $E = \sum_{n=0}^{N-1} X(n)e^{-jx(n)k}$   
 $E = \sum_{n=0}^{N-1} X(n)e^{-jx(n)k}$   
 $E = \sum_{n=0}^{N-1} X(k)e^{-jx(n)k}$   
 $E = \sum_{n=0}^{N-1} X(k)e^{-jx(n)k}$ 

FFT (Fast Fourier Transform): \* It is an algoritam weal to compute the. DFT. \* It reduces the computation time required to compute discrete fourier transform. Types of FFT:-1. Decimation-in-time (DIT) 2. Decimation-in-frequency (DIF) 2. Decimation-in-frequency (DIF) 3. For DIT the input is bit revosal while the output \* for DIT the input is bit revosal while the output is in natural order while the output is bit reversal. is natural order while the output is bit reversal. . Decimation-in-time (DIT)

Bit reversal.

Docimal	Binary	Bit-reversal Order	Bit-revexal cindea
0	000	000	10)
	001	100	4 even
2	010	010	2
2	011	110	65
	100	001	17
4	101	101	5 204
5	110	011	3
6	111		170

TETT DFT properties are divided to 4-points. step 1: output "-The B(0) = A(0)+A(4) B(1) = A(1)+A(5) XID A(O) B(O) X(0). P(n) Rio B(2) = A(2)+A(6) X(4) 4-point X(I) P (2) B(2) X(2) X(2)-B(3) = A(3)+A(7) JFT B(3) ACE (N/2)  $\chi(3)$ x(6) B(4)=B(0)-A(4) B(5) - A(1) - A(5) A(4) X(I) B(4) X(4) B(6) - P(2) - P (6) A-Point A(S) B(5) x(5) x(5) B(7)= B(3)-A(7) A(6) BB DFT X(6)  $\chi(3)$ A(7) B(7) (N/2) X(7) (7) are divided into 2-points. step2:-DFT properties 9(0) X(0) The Flo) Blo) A(O) <u>X(1)</u> FU B(1) X (0) /gci) NA ACI) DFT X(H) 2-point g(2) X (2) B(2) Q R2) Wo (3)X(3)  $\gamma(2)$ (FC) <u>}B(3)</u> N/4 A(2) W& A(3) DFT-2 2-point 210 -1 9(4) X(4) WS B(4) A(4) <u>X(5</u>) N/4 DFT-3 P(5) 2(1) -1 g(5) Wg 7(5) BS) 2-point AG 9(6) X(6) W8 B(E) Ø W8 A14) X(3) N/4 X(7) F(G) (BCI) AT (r)x DFT-4 W8 F(9) g(7) 2. point Wg -1 Step 3:-Twiddle factor WNK = e-jett KN  $W_8^6 = e^6 = 1$  $W_8^6 = 1$ when K=D, N=8 /

Step 4:-  
The definition of DFT is,  

$$N-1$$
  $N-1$   $X(R) = \sum_{n=0}^{N-1} \chi(n) e^{-j a T (n) K N}$ 

For even values,  

$$\chi(n) = \chi(an)$$
  
 $\chi(k) = \underset{h=0}{\overset{N-1}{\approx}} \chi(an)e^{-jannk}N$   
For add values,

odd Values,  

$$\chi(n) = \chi(2n+1)$$
  
 $\chi(z) = \sum_{n=odd} \chi(2n+1) e^{-j a \pi n k} N$ 

1.25 steps .a-tbWNK 0 K a- bw WNK Stepb:-Butterfly structure for DIT is, Complete ro) -X(0)XID (4) - X(2)  $\chi(Q)$ X(3) (b) W. 入(4) 7/1 XUS Wg', x(0) W82 1)× x(6)-ょい XTT Wg stage 3 Stage 2 stage 1 DFT of a sequence xcn)={1,2,3,4,4,3,2,13 EX.1.10. Compute Huning DIT-FFT algorithm. 6 Step 1: Twiddle factor:-WN = e-jatik/N when k=0,  $W_8^0 = e^0 = 1$ (DX N=8,  $W_8^0 = 1$ 

when k=1, W8' = e-jaT1/8 W8 - 0.707 - jo.707 When k=2,  $W_{g}^{2} = e^{-j2\pi(2)}$  $W_8^2 = -j$ When k=3,  $W_8^3 = e^{-\int \frac{2\pi}{8} (3)}$   $W_8^3 = -\int \frac{2\pi}{8} (3)$   $W_8^3 = -\int \frac{2\pi}{8} (3)$   $W_8^3 = -\int \frac{2\pi}{8} (3)$  $\chi(1) = 2, \chi(2) = 3, \chi(3) = 4, \chi(4) = 4$ Ilp sequence:-ス(1)=マ, X ス(5)=3, ス(6)=2, ス(7)こし メ(の) X(K) 20 X(0) 5+5=10 10 1+4=5  $\chi(0) = 1$ ร -5.828-j2.414 X(1) 1-4=3 x(4)= @ .4 -2--3 X(2) X(Q)= 3Wg=1 3+2=5 1 5 5-5=0 0 0 0.110-j0:414 J<sup>3tj</sup> x(B)=\$\$ -31) 3-2=12-1-X(3) x(4)=de w XLA 0 545=10 2+3=5 -0.172 + 0.4 kg N(5)=3 2-3=-1 -2.828-1.44 0 x (1)  $\chi(B)=4$ 5-5-0 4+1=510-1 5 0 -5.808+j2414 1-1+3j 2.808-j1,414 4-1=3 -1 n(7)=1 XIN 3] ו- רסריםן רסה ס-א X(K)= {20, -5.828-j2.414,0,0.172-j0.414,0,-0.172+j0.414,0, 0, -5.821

5.1.11 Determine 4 - point DFT of a required 
$$\chi(n) = \{1, 2, 5, 4\}$$
  
Wing DIT FFT algorithm.  
 $\chi(n) = \{1, 2, 3, 4\}$   
Twieddle factor.  
 $N = 4$ ,  $W_{h}^{k} = e^{-j \otimes T \frac{k}{N}}$   
 $K = 0$ ,  $W_{h}^{0} = e^{0} = 1 \Rightarrow W_{h}^{0} = 1$ ,  $W_{h}^{0} = 1$ ,  $W_{h}^{0} = 1$ ,  $W_{h}^{0} = e^{-j \otimes T \frac{k}{N}}$   
 $k = 1$ ,  $W_{h}^{1} = e^{-j \otimes T \frac{k}{N}} = e^{-j \frac{N}{2} \frac{k}{2}}$ ,  $W_{h}^{0}$   
 $\chi(n)$   
 $\chi(n)$   
 $\chi(n)$   
 $\chi(n) = 1$   
 $\chi(n) = 2$   
 $\chi(1) = 2$   
 $\chi(3) = 4$   
 $\chi(3) = -2$   
 $\chi(3) = 4$   
 $\chi(k) = \{10, -2+2j, -2, -2-2j\}$   
 $\chi(k) = \{10, -2+2j, -2, -2-2j\}$   
 $\chi(k) = \{10, -2+2j, -2, -2-2j\}$   
DIF (Decimation in Frequency)  
 $\beta = 1$ : Divide DFT by N/4  
 $N = 8$ 

0



Output:- B(0) = B(0) + B(4) B(1) = B(5) + B(1) B(2) = A(2) + A(6) B(3) = A(3) + B(7) B(4) = B(0) - B(4) B(5) = A(1) - B(5) B(6) = B(2) - B(6)B(7) = B(3) - B(7)



of  $\chi(n)$  is  $-j \partial \pi h(n) e$   $\chi(k) = \chi(n) e$ DFT for odd values, x(n)=x(an+1)  $X(k) = \sum_{h=0}^{N-1} \chi(a_{h+1})e^{-ja_{11}k/N}$ for even values x(n) = x(an)  $X(k) = \sum_{n=0}^{N-1} \chi(an) e^{-j QTT k} / N$ Twiddle factor. N=8, WNK = e-jaTIK/N k=0, WR = e=1 Mg = 1 k=1,  $w_{s}^{1} = e^{-j \vartheta T_{s}} = e^{-j \Re t_{s}}$ Wg = 0.707-j0.707 k = 2,  $W_{q}^{2} = 0$ , j = 0,  $\frac{310 \times 2}{8} = 0$ , j = 0 $w_{g}^{2} = -j$ k = 3,  $W_8^3 = e^{-j\frac{2\pi i^3}{8}} = e^{-j\frac{\pi i^3}{4}}$  $W_8^3 = -0.707 \cdot 0^{-0.707}$ Basic Butter-Sly Diagram:a+b ā WNK [a-DWN Ŀ

Complete Buttorfly structure for DIF. X6) X(0) W8 X(4)  $\mathcal{N}(\mathcal{I})$ X(2) 7(2) W8 W8 X(6) NB) W82 X(I)X[4) W80  $\sim W_{s}^{o}$  X(s)  $\chi(s)$ W W8 X (3) <u>X (6)</u> W8 <u>√ W8°</u> ×(7) X(7) e Wg ₩<sup>3</sup>



N=8- n=0,1,2,3,4,5,16,7  $\chi(0)=1, \chi(1)=2, \chi(2)=3, \chi(3)=4, \chi(5)=6$   $\chi(6)=7, \chi(7)=8.$ Twieddle factor:- $W_{N}^{K}=e^{-jRTi\frac{K}{N}}$   $W_{N}^{K}=e^{-jRTi\frac{K}{N}}$   $W_{8}^{R}=1, \quad W_{8}^{R}=-0.707-j0.707$   $W_{8}^{R}=-j, \quad W_{8}^{R}=-0.707-j0.707$ 



EX. 1.B. Datermine 4-point DFT of a sequence  

$$x(n) = \{1, 2, 3, 4\}$$
 wring DIF - FFT algorithm  
 $x(n) = \{1, 2, 3, 4\}$  wring  $x(n) = -FFT$  algorithm  
 $x(n) = \{1, 2, 3, 4\}$  wring  $x(n) = -FFT$  algorithm  
 $x(n) = \{1, 2, 3, 4\}$   
 $w_{N} = \{1, 2, 3, 4\}$  wring  $x(n) = -FFT$  algorithm  
 $x(n) = \{1, 2, 3, 4\}$   
 $w_{N} = \{2, 3, 4\}$   



= (U)X

1.33





 $x_{(n)} : \{ x_{(n)}, y_{(n)}, y_{(n)}, y_{(n)}, y_{(n)} \}$ 

Discrete - Time Fourier transform -

The DIFT of discrete-time signal xin is,

$$X(e^{j\omega}) = \underset{n=-\infty}{\overset{\forall}{\underset{n=-\infty}{x(n)e}}} x(n)e^{-j\omega n}$$

The IDTFT of  $X(e^{j\omega})$  is, TT $X(n) = \frac{1}{8TT} \int X(e^{j\omega})e^{j\omega n} d\omega$ 

Find the fourier transform of the following. 9 S(n)

$$\chi(m) = \delta(n)$$

$$\chi(e^{j\omega}) = \underbrace{\Xi}_{n=-\infty}^{\infty} \chi(n) e^{-j\omega n}$$

$$= \underbrace{\Xi}_{n=-\infty}^{\infty} \delta(n) e^{-j\omega n}$$

$$\int (n) = 0 \quad \text{for } n \neq 0$$

$$\int for \quad n = 0$$

b)  $\chi(m) = u(n)$   $\chi(e^{ju}) = \sum_{n=-\infty}^{\infty} \chi(m)e^{-jun}$   $\chi(e^{ju}) = \sum_{n=0}^{\infty} e^{-jun} = 1$  $\chi(e^{ju}) = 1 + e^{-ju} + e^{-ju} + \cdots + \infty$ 



$$u(n) = 0 \quad \text{for } n = 0$$

$$I \quad \text{for } n \ge 0$$

$$I + \gamma + \gamma^2 + \cdots \ll = \frac{1}{1 - \gamma}$$

$$\chi(n) = \delta(n-k)$$

$$X(e^{jw}) = \sum_{n=-a}^{\infty} \chi(n)e^{-jwn}$$

$$= \sum_{n=0}^{\infty} \delta(n-k)e^{-jwn}$$

$$\chi(e^{jw}) = e^{-jwk}$$

$$\delta(n-k) = 1 \quad \text{for } n=k$$

$$= 0 \quad \text{for } n \neq k$$

$$\begin{aligned} \chi(n) &= u(n-k) \\ \chi(e^{j\omega}) &= \stackrel{\forall}{\underset{n=-\infty}{\overset{\forall}{\overset{\forall}{n=-\infty}}}} u(n-k) e^{-j\omega n} \\ &= \stackrel{=}{\underset{n=k}{\overset{\forall}{e}}} e^{-j\omega n} \\ &= e^{-j\omega k} + e^{-j\omega (k+1)} + \cdots \\ &= e^{-j\omega k} (1 + e^{-j\omega} + e^{-j2\omega} + \cdots) \\ &= e^{-j\omega k} (1 + e^{-j\omega} + e^{-j2\omega} + \cdots) \\ &= e^{-j\omega k} (1 - e^{-j\omega} + e^{-j\omega} + \cdots) \\ &= e^{-j\omega k} (1 - e^{-j\omega} + e^{-j\omega} + \cdots) \end{aligned}$$

Derivation of DFT from DIFT: one period  
The DFT of X(n) is obtained by sampling, DTFT  

$$X(e^{jw})$$
 at a finite number of frequency  
Points:  
The sampling frequency points are,  
 $w_k = \frac{2\pi k}{N}$ ; for  $K:0, b:0, \dots, N-1$   
where  $N \rightarrow total$  no. of samples,  
The sampling of  $X(e^{jw})$  is mathematically  
expressed as,  
 $X(K) = X(e^{jw})/w = \frac{0\pi k}{N}$ ; for  $k:91, 0, \dots, N-1$   
 $ie)$   $X(K) = \frac{N-1}{0=0}$   $Y(n) e^{-j\frac{\pi}{N}}$ ; for  $k:91, 0, \dots, N-1$   
 $ie)$   $X(K) = \{X(0), X(1), X(2), \dots, -X(N-1)\}$   
The Inverse DFT of  $X(k)$  of length  $N$  is,  
 $TDFT(X(k)) = Y(0) = \frac{N-1}{N}$   $X(k) e^{j\frac{\pi}{N}}$ ; for  $n = 0, 1, N-1$ 

D compute DFT of xcm= g1, 1, -2, -23 Ans: {-2, 3-3j, 0, 3+3j 2) Find IDFT of Y(K)= {1,0,60} y(n)= 1 E Y(x)e n=0, 1, - N-1 Ans: [0.5,0,0.5,0] 3) find DFT of \$1,1,0,03 Ans: X(K) = { a, 1-j, 0, 1+j] 4) DFT of (-1) for N=4  $\chi(n) = (-1)^{h}$  $\chi(0) = 1$ ,  $\chi(1) = -1$ ,  $\chi(2) = 1$ ,  $\chi(3) = -1$ Ans: X(K) = \$0,0, 4,03 5) comput 4-point DI-T of the sequence  $\mathcal{H}(n) = 6 + Sin\left(\frac{a \pi n}{N}\right) = n = 0, 1, \dots N^{-1}$ X(0)=6, X(n=7, X(0)=6, X(3)=5 Ans! X(K) = {84, -je, 0, je?

Ex.1.

Find IDFT of the sequence.  $X(K) = \{5, 0, 1-j, 0, 1, 0, 1+j, 0\}$  $X(n) = \frac{N^{-1}}{N} \frac{X(K)}{K=0} \times (K) e^{j e^{i \pi i K n/N}} n=0, 1, \dots, N-1$ 

for N=8,  

$$N-1$$
  $\chi(k) = \frac{1}{8} \sum_{k=0}^{N-1} \chi(k) = \frac{1}{8} \sum_{k=0}^{N-1$ 

For n:0,  

$$\chi(0) = \frac{1}{8} \left[ \frac{7}{k=0} \chi(k) \right]$$
  
 $= \frac{1}{8} \left[ 5+0+1-j+0+1+0+1+j+0 \right]$ 

$$\begin{aligned} \chi(0) &= 1 \\ \chi(1) &= \frac{1}{8} \left[ \sum_{k=0}^{7} \chi(k) e^{j\pi k/4} \right] \\ &= \frac{1}{8} \left[ \frac{1}{8} + (1-j) + (1-j) + (1-j) + (1-j) \right] \end{aligned}$$

$$\begin{aligned} &= \frac{1}{8} \begin{bmatrix} 6 \end{bmatrix} \\ \chi(n) &= 0.75 \\ J \\ = \frac{1}{8} \begin{bmatrix} -\frac{1}{2} \\ K=0 \end{bmatrix} \\ K=0 \end{bmatrix} \\ &= \frac{1}{8} \begin{bmatrix} -\frac{1}{2} \\ K=0 \end{bmatrix} \\ &= \frac{1}{8} \begin{bmatrix} 5 + (1-j)(-i) + [Xi] + (1+j)(-i) \end{bmatrix} \\ &= \frac{1}{8} \begin{bmatrix} 5 + (1-j)(-i) + [Xi] + (1+j)(-i) \end{bmatrix} \\ \chi(n) &= \frac{1}{8} \begin{bmatrix} 4 \end{bmatrix} = 0.5 \end{aligned}$$

$$\chi(3) = \frac{1}{8} \left[ \sum_{k=0}^{7} \chi(k) e^{j 3\pi k/4} \right]$$
  
=  $\frac{1}{8} \left[ \pi + (1-j)(-j) + 1(-1) + (1+j)(j) \right]$   
=  $\frac{1}{8} \left[ 2 \right]$   
 $\chi(5) = 0.85$   
 $\chi(4) = \frac{1}{8} \left[ \sum_{k=0}^{7} \chi(k) e^{j 5\pi k/4} \right]$ 

$$= 1$$
  

$$\gamma(s) = \frac{1}{8} \left[ \frac{1}{2k} \times (k) e^{j \cdot 5\pi i k/4} \right]$$
  

$$= \frac{1}{8} \left[ 5 + (1 - j) j + (1)(-1) + (1 + j)(-j) \right]$$
  

$$= \frac{1}{8} \left[ 6 \right] = 0.75$$
  

$$\chi(b) = \frac{1}{8} \left[ \frac{1}{2k} \times (k) e^{j \cdot 3\pi i k/8} \right]$$
  

$$= \frac{1}{8} \left[ 4 \right] = 0.5$$
  

$$\chi(\tau) = \frac{1}{8} \left[ \frac{1}{2k} \times (k) e^{j \cdot 7\pi k/4} \right]$$
  

$$= \frac{1}{8} \left[ 4 \right] = 0.5$$
  

$$\chi(\tau) = \frac{1}{8} \left[ \frac{1}{2k} \times (k) e^{j \cdot 7\pi k/4} \right]$$
  

$$= \frac{1}{8} \left[ 6 + (1 - j)(-j) + 1(-1) + (1 + j)(j) \right]$$
  

$$= \frac{1}{8} (a) = 0.85$$
  

$$\chi(c_{0}) = .\frac{5}{8} 1, 0.75, 0.5, 0.85, 1, 0.75, 0.50.85$$

\*  $\leq r$ . Determine the 8-point DFT of the sequence:  $\chi(n) = \beta_{1,1,1,1,1,1,0,0}$  $\chi(k) = \sum_{n=0}^{N-1} \chi(n) e^{-ja_{1,1}nk/N}$  $\chi(k) = \sum_{n=0}^{N-1} \chi(n) e^{-ja_{1,1}nk/N}$ 

> for N=8,  $X(k) = \prod_{n=0}^{3} \chi(n) e^{-j \pi n k/4} j k = 0, 1, ...7$

1) obtain the circular convolution of the following Sequences.  $\chi(m) = \{1, 2, 1\}, h(m) = \{1, -2, 2\}$  $\begin{cases} 1 & | & 2 \\ 2 & | & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ | & 2 & | \\ |$ Y(b)= {3, 2, -13 Linear convolution:no. of . Samples in the output is N = L+M-1 Circular convolution:--No. of. Samples in the output N= man(L,M) 1. Compute linear & circular convolution of two H.W: sequences. 2, (n) < \$1,1,2,2} and xe(n)= \$1,8,3,4} Ans: - for circular X3(10)= {15,17,15,13} ) Find the linear convolution of  $x(n) \cdot \{1, -3, 5, -7, 9, -1\}$ with h(s)= {-4, 8, -16) using circular convolution. NOD: { 1, -3, 5, -7, 9, -113 ie) 1=6 h(n)= {-4,8,-163 ie) M=3 N= L+M-1=6+3-1 N=8

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wing overtap add method.

ip Periodicity:

If a sequence acri is periodic with Periodicity of N Samples than N-point DFT of the sequence is also periodic with the Periodicity of n samples. x(n+N) = x(n); for all values of n Proof:- X(K+N) = X(K); for all values of K. By definition of IDFT,  $\chi(n) = \frac{1}{N} \sum_{k=0}^{N-1} \chi(k) e^{j \frac{2\Pi k n}{N}} , n = 0, 1, \dots - 1$  $\sum_{k=0}^{N-1} \chi(n+N) = \frac{1}{N} \sum_{K=0}^{N-1} \chi(K) = \frac{1}{N} \sum_{K=0}^{N-1} \chi(K) = \frac{1}{N}$ Put n= n+N  $Y(n+N) = \frac{1}{N} \underset{k=0}{\overset{N-1}{\underset{k=0}{\times}} \chi(k)} e^{j\frac{2\pi i k n}{N}} e^{j\frac{2\pi i k n}{N}}$  $= \frac{1}{N} \sum_{k=0}^{N-1} \chi(k) e^{j a \pi k} e^{j a \pi k}$ 2=1  $\begin{array}{ccc} \kappa=0 & N-1 \\ \chi(n+N) & = 1 \\ N & \kappa=0 \end{array} \begin{array}{c} j & j & \overline{\lambda} & n \\ \overline{N} & \left( e^{j & n & \kappa} & e^{j & n & \kappa} \right) \end{array}$  $\chi(n+N) = \chi(n)$ 

By the definition of DFT,  

$$X(k) = \sum_{n=0}^{N+1} \chi(n) e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}} \cdot k=0, \dots N-1}$$

$$p_{n} = k = k + N,$$

$$Y(k + N) = \sum_{n=0}^{N-1} \chi(n) e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}} \cdot e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}}$$

$$= \sum_{n=0}^{N-1} \chi(n) e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}} \cdot e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}}$$

$$X(k + N) = \chi(k)$$

$$Herio proved.$$
(ii) Circular Convolution:  
The DFT of a Circular convolution of  
two Sequences in time domain  

$$I = DFT [\chi_{1}(n)] = \chi_{1}(k) + M$$

$$DFT [\chi_{1}(n)] = \chi_{0}(k) + Ren$$

$$DFT [\chi_{1}(n)] = \chi_{0}(k) + Ren$$

$$DFT [\chi_{1}(n)] = \sum_{n=0}^{N-1} \chi(n) e^{-\int_{0}^{2\pi} \frac{\pi}{N} \frac{1}{N}}$$

$$\chi_{i}(m) * \chi_{q}(n) = \sum_{n=0}^{N-1} \chi_{i}(m) \cdot \chi_{q}(n-m)$$
  
DFT[ $\chi_{i}(n) * \chi_{q}(n)$ ] =  $\sum_{n=0}^{N-1} \sum_{m=0}^{N-1} \chi_{i}(m) \cdot \chi_{q}(n-m) \cdot O^{-1} \cdot \sum_{n=0}^{N-1} \chi_{i}(m) \cdot \chi_{q}(m) \cdot \chi_{q}(m$ 

$$\begin{split} \mathcal{L} &= n - m \\ n &= m + l \\ &= \sum_{m=0}^{N-1} \pi_1(m) \sum_{l=0}^{N-1} \chi_0(l) e^{-j \alpha T T} (m + l) K N \\ &= \sum_{m=0}^{N-1} \pi_1(m) \sum_{l=0}^{N-1} \chi_0(l) e^{-j \alpha T T} m K N \\ &= \sum_{m=0}^{N-1} \pi_1(m) \sum_{l=0}^{N-1} \chi_0(l) e^{-j \alpha T T} m K N \\ &= \sum_{m=0}^{N-1} \pi_1(m) e^{-j \alpha T T} m K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K N \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K \\ &= \sum_{m=0}^{N-1} \chi_1(m) e^{-j \alpha T} \pi K \\ &= \sum_{m=0}^{N-1}$$

$$DFT[m,(m*M_{Q}(m)] = X_1(K) X_Q(X)$$

Symmetry:  
If 
$$x(n)$$
 is real,  
 $X(k) = x^* (N-k)$   
 $X_{R}(k) = X_{R}(N-k)$   
 $X_{f}(k) = -X_{f}(N-k)$   
 $|X(k)| = -X_{f}(N-k)|$   
 $|X(k)| = |X(N-k)|$   
 $|X(k)| = -(X(N-k))|$   
 $|X(k)| = -$ 

$$= \left( \sum_{k=0}^{N-1} \chi_{(n)} e^{j \# \Pi kn/N} \right)^{*}$$

$$= \left[ \sum_{k=0}^{N-1} \chi_{(n)} e^{j \# \Pi kn/N} - j \# \Pi Nn/N} \right]^{*}$$

$$= \left[ \sum_{k=0}^{N-1} \chi_{(n)} e^{-j \# \Pi (N-k)n/N} \right]^{*}$$

$$\times (k) = \cdot \chi^{*} (N-k)$$

$$\cdot \cdot \chi(k) = \chi^{*} (N-k)$$

$$\chi(k) = \chi^{*} (N-k)$$

$$= \left[ \chi_{R}(N-k) + j \chi_{T}(k) \right]^{*}$$

$$= \left[ \chi_{R}(N-k) + j \chi_{T}(k) - k \right]^{*}$$

$$\chi(k) = \chi_{R}(k-k) - j \chi_{T}(N-k) \right]^{*}$$

$$\chi(k) = -\chi_{R}(k-k)$$

$$\chi_{R}(k) = \chi_{R}(N-k)$$

$$\chi_{R}(k) = -\chi_{T}(N-k)$$

$$\chi_{R}(k) = -\chi_{T}(N-k)$$

$$\chi_{R}(k) = \sqrt{\chi_{R}(N-k)}^{*} + \chi_{T}(k)^{2} = \sqrt{\chi_{R}(n-k)}^{*} + \chi_{T}(n-k)^{2}$$

$$[\chi(N-k)] = \left[ \chi_{R}(N-k)^{2} + \chi_{T}(k)^{2} = \sqrt{\chi_{R}(n-k)} + \chi_{T}(n-k)^{2} + \chi_{T}(k-k)^{2} + \chi_{T}(k) + \chi_{T}(k-k)^{2} + \chi_{T}(k-k)^{2} + \chi_{T}(k) + \chi_{T}(k-k)^{2}$$

$$[\chi(N-k)] = \left[ \chi_{R}(N-k)^{2} + \chi_{T}(k) + \chi_{T}(k-k) + \chi_{T}(k) + \chi_{T}(k-k) + \chi_{T}(k) + \chi_{T}(k) + \chi_{T}(k-k) + \chi_{T}(k-k) + \chi_{T}(k) + \chi_{T}(k-k) + \chi_{T}(k-k) + \chi_{T}(k) +$$

St. C.